Optimization in Wireless Multi-relay Networks

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Outline

- Introduction to Cooperative Communication
 - Cooperative Communication
- Part I: Distributed Space-Time Coding (DSTC)
 - System Model
 - Fourier-based DUSTM
 - Power Allocation (PA) in DSTC
 - Optimal Training and Mismatched Decoding in DSTC
- 3 Part II: Distributed Beamforming
 - Introduction & System Model
 - Guaranteed QoS
 - SNR Margin Maximization

Conclusion

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Introduction to Cooperative Communication Part I: Distributed Space-Time Coding (DSTC)

Part I: Distributed Space-Time Coding (DSTC) Part II: Distributed Beamforming Conclusion Cooperative Communication

Cooperative Communication - Overview



- A new form of spatial diversity.
- Users cooperate to relay signals of each other, and emulate a virtual array of transmit antennas.
- Huge potential in improving the reliability of the wireless network.



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Introduction to Cooperative Communication

Part I: Distributed Space-Time Coding (DSTC) Part II: Distributed Beamforming Conclusion

Cooperative Communication

Amplify-and-forward - Two stages of transmission



- First stage: Source (S) transmits to both Relay (R) and Destination (D).
- Second stage: Relay amplifies the received signal and forwards it to Destination.
- Destination combines the two received signals to decode.



System Model Fourier-based DUSTM Power Allocation (PA) in DSTC

System Model (in Part I)



- One antenna per node, used for both TX and RX.
- R relays work in half-duplex mode, Amplify-and-Forward (AF) protocol is considered.
- No direct link from source to destination.

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Distributed Space-Time Coding (DSTC)





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Distributed Space-Time Coding (DSTC)





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Distributed Space-Time Coding (DSTC) (cont.)

• The mathematical model is

$$\mathbf{y} = \sum_{i=1}^{R} \tilde{g}_i \mathbf{t}_i + \mathbf{z}_D = \mathbf{X} \mathbf{\Lambda} \mathbf{h} + \mathbf{z}$$

where $\mathbf{X} = [\mathbf{A}_{1}\mathbf{s}, \dots, \mathbf{A}_{R}\mathbf{s}]$ $\mathbf{\Lambda} = \operatorname{diag}\left(\sqrt{\varepsilon_{1}\sigma_{F_{1}}^{2}\sigma_{G_{1}}^{2}}, \dots, \sqrt{\varepsilon_{R}\sigma_{F_{R}}^{2}\sigma_{G_{R}}^{2}}\right)$ $\mathbf{h} = \left[f_{1}^{(*)}g_{1}, \dots, f_{R}^{(*)}g_{R}\right]^{T}$ $\mathbf{z} = \frac{1}{\sqrt{P_{0}T\sigma_{R}^{2}}}\sum_{i=1}^{R}\sqrt{\varepsilon_{i}\sigma_{G_{i}}^{2}}g_{i}\mathbf{A}_{i}\mathbf{z}_{R_{i}}^{(*)} + \frac{1}{\sqrt{P_{0}T\sigma_{D}^{2}}}\mathbf{z}_{D}.$

• X is now a distributed space-time codeword.



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Early Works and My Works in DSTC

- Applying ST coding in MIMO systems to relay networks.
- Coherent networks (CSI of S → R, R → D known): linear dispersion DSTC [Jing06], orthogonal DSTC [Jing07].
- Partially coherent networks (only CSI of R → D known): differential DSTC [Kiran07].
- Noncoherent networks (CSI unknown): cyclic DSTC [Oggier06].
- My approach:
 - Propose Fourier-based Distributed Unitary Space-Time Modulation (DUSTM): design source signal s_k and relaying matrix A_i → design X_k.
 - Provide a unified analysis for partially-coherent and noncoherent networks.



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DUSTM in a Partially Coherent Network

The ML decoding is to maximize the probability p(y|X_k, {g_i}). The decoding rule could be found as

$$\mathbf{X}_{ML} = \arg \max_{\mathbf{X}_k = \mathbf{X}_1, \dots, \mathbf{X}_L} \mathbf{y}^{\mathcal{H}} \mathbf{X}_k \mathbf{C} \mathbf{X}_k^{\mathcal{H}} \mathbf{y}, \qquad (1)$$

where

$$\mathbf{C} = \operatorname{diag}\left(\frac{\beta_1|g_1|^2}{\gamma + \beta_1|g_1|^2}, \dots, \frac{\beta_R|g_R|^2}{\gamma + \beta_R|g_R|^2}\right)$$
$$\gamma = \frac{1}{P_0 T} \left(1 + \sum_{i=1}^R \varepsilon_i \sigma_{G_i}^2 |g_i|^2\right)$$
$$\beta_i = \varepsilon_i \sigma_{F_i}^2 \sigma_{G_i}^2.$$



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DUSTM in a Noncoherent Network

- p(y|X_k) does not appear to have a closed-form expression. The optimal ML decoder is *unavailable*.
- A suboptimal Generalized Likelihood Ratio Test (GLRT) decoder can be derived as

$$\mathbf{X}_{GRTL} = \arg \max_{\mathbf{X}_k = \mathbf{X}_1, \dots, \mathbf{X}_L} \mathbf{y}^{\mathcal{H}} \mathbf{X}_k \mathbf{X}_k^{\mathcal{H}} \mathbf{y}.$$
 (2)

• *Remarks:* The difference between the ML decoder in (1) and the GLRT decoder in (2) is the existence of the matrix C, which contains the CSI of the $R \rightarrow D$ channels.



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Partially-coherent vs. Noncoherent



• Symbol error performance of DUSTM: $\sigma_{F_i}^2 = 10$ and $\sigma_{G_i}^2 = 1$.

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The Optimization Problem

• The average SNR at the destination:

$$SNR = \frac{P_0 T}{RN_0} \frac{\sum_{i=1}^R \varepsilon_i \sigma_{F_i}^2 \sigma_{G_i}^2}{1 + \sum_{i=1}^R \varepsilon_i \sigma_{G_i}^2}.$$
 (3)

• The optimization problem



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Balanced Networks

- Early works *only* consider the special case: $\sigma_{F_i}^2 = \sigma_{G_i}^2 = 1$, the "Equal PA" scheme $P_0 = P/2$, $P_i = P/(2R)$ is optimal.
- My approach: Study the PA scheme for arbitrary $\sigma_{F_i}^2$ and $\sigma_{G_i}^2$.
- Balanced networks: $\sigma_{F_1}^2 = \ldots = \sigma_{F_R}^2 = \sigma_F^2$, and $\sigma_{G_1}^2 = \ldots = \sigma_{G_R}^2 = \sigma_G^2$, the optimal PA scheme is

$$P_{0} = \begin{cases} \frac{\sqrt{(P\sigma_{F}^{2}+N_{0})(P\sigma_{G}^{2}+N_{0})}-(P\sigma_{G}^{2}+N_{0})}{\sigma_{F}^{2}-\sigma_{G}^{2}}, \text{ if } \sigma_{F}^{2} \neq \sigma_{G}^{2} \\ P/2, & \text{ if } \sigma_{F}^{2} = \sigma_{G}^{2} \\ P_{1} = \dots = P_{R} = (P-P_{0})/R. \end{cases}$$



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Unbalanced Networks

- The network is called unbalanced if the conditions $\sigma_{F_1}^2 = \ldots = \sigma_{F_R}^2$ and $\sigma_{G_1}^2 = \ldots = \sigma_{G_R}^2$ are not met.
- To get the maximum SNR, the relay power is allocated to the best relay, say the *j*th relay. Thus, only one fading path is active → compromise the performance of the DSTC.



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Balancing the Unbalanced Networks

- The amount of fading (AoF) is a common measure of fading severity in a fading channel model.
- Establish the condition to minimize the amount of fading in a relay network to balance the fading statistics of each $S \rightarrow R \rightarrow D$ link.
- With the AoF constraint, the optimal PA scheme

$$P_{0} = \begin{cases} \frac{\sqrt{(Pa+c)(Pb+c)} - (Pa+c)}{b-a}, & \text{if } b \neq a\\ P/2, & \text{if } b = a \end{cases}$$
$$P_{i} = \frac{P - P_{0}}{P_{0}b + c} \cdot \frac{P_{0}\sigma_{F_{i}}^{2} + N_{0}}{\sigma_{F_{i}}^{2}\sigma_{G_{i}}^{2}}, \quad i = 1, \dots, R, \qquad (4)$$

where a, b, and c are parameters.

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Some Remarks

- The proposed PA scheme achieves the maximum diversity order in all coherent, partially coherent, and noncoherent relay networks.
- The proposed PA scheme yields a significant performance advantage over the "equal PA" scheme.

• Consider the unbalanced network:



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Noncoherent Network



 Distributed Unitary Space-Time Modulation (USTM) is applied to the noncoherent network with 2 and 3 relays.



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Coherent Network



 Distributed Orthogonal Space-Time Block Coding (OSTBC) is applied to the coherent network with 2 and 4 relays.

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Optimal Training in DSTC

- Coherent ML decoding of DSTC requires the knowledge of channel coefficient vector **h**.
- Need to estimate **h** at the destination:
 - $\bullet\,$ Send a known training sequence u from the source

$$\mathbf{y}_{\mathcal{T}} = \mathbf{X}_{\mathcal{T}} \mathbf{\Lambda} \mathbf{h} + \mathbf{z}_{\mathcal{T}},$$

where
$$\mathbf{X}_T = [\mathbf{A}_1 \mathbf{u}^{(*)}, \dots, \mathbf{A}_R \mathbf{u}^{(*)}].$$

- Estimate **h** from \mathbf{y}_T , \mathbf{X}_T , and $\mathbf{\Lambda}$.
- Early works studies the optimal design of X_T [Gao08].
- My approach: find the optimal PA scheme in training phase, and investigate the impact of imperfect CSI to the coherent code.



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Mean-Square Error in Channel Estimation

• Maximum Likelihood (ML) estimation:

$$\operatorname{cov}\left(\boldsymbol{\Delta}_{h}\right)=\bar{\gamma}\boldsymbol{\Lambda}^{-1}(\boldsymbol{X}_{T}^{\mathcal{H}}\boldsymbol{X}_{T})^{-1}\boldsymbol{\Lambda}^{-1}.$$

where
$$\bar{\gamma} = \mathbb{E}[\gamma] = \frac{1}{P_0 T} \left(1 + \sum_{i=1}^R \varepsilon_i \sigma_{G_i}^2 |g_i|^2 \right)$$

• Minimum Mean-Square Error (MMSE) estimation

$$\operatorname{cov}(\mathbf{\Delta}_h) = \left(\mathbf{I}_R + \frac{1}{\bar{\gamma}}\mathbf{\Lambda}\mathbf{X}_T^{\mathcal{H}}\mathbf{X}_T\mathbf{\Lambda}\right)^{-1}$$

- The mean-square error (MSE) is minimized when
 - X_T is orthogonal [Gao08].
 - The optimal PA scheme with the minimum amount of fading constraint.



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Coherent Network



 Total MSE achieved with ML and MMSE estimators, and with the optimal and equal PA schemes.



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Mismatched Decoding

• Recall the system model

$$\mathbf{y} = \mathbf{X} \mathbf{\Lambda} \mathbf{h} + \mathbf{z}.$$

• Use the estimated CSI $\hat{\mathbf{h}}$

$$\hat{\mathbf{X}} = \arg\min_{\mathbf{X}_k} \|\mathbf{y} - \mathbf{X}_k \mathbf{\Lambda} \hat{\mathbf{h}} \|^2.$$

• The same diversity order is achieved with imperfect CSI estimation as with perfect CSI.



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System Model Fourier-based DUSTM Power Allocation (PA) in DSTC Optimal Training and Mismatched Decoding in DSTC

Coherent Network



Error performance of DSTC with different types of decoding.
Red lines: Optimal PA, Blue lines: Equal PA.



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Introduction & System Model Guaranteed QoS SNR Margin Maximization

Introduction to Part II

- With full CSI, the relays can beam the retransmitted signal to the destination ⇒ received signal is coherently constructed.
- Early works *only* consider power allocation at the relays for a one-source one-destination network.
- My approach: Find optimal power allocation for a multiple-source multiple-destination network
 - (i) Minimizing the sum relay power with guaranteed quality of service (QoS).
 - (ii) Maximizing the joint SNR margin subject to power constraints at the relays.
 - Apply convex optimization to investigate the problems.
 - Propose simple and fast converging algorithms.



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Introduction & System Model Guaranteed QoS SNR Margin Maximization

System Model (in Part II)



Relay R

• N users $(S_n - D_n)$ compete for the power resource at the R relays.



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Introduction & System Model Guaranteed QoS SNR Margin Maximization

System Model

• First stage





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Introduction & System Model Guaranteed QoS SNR Margin Maximization

System Model

First stage



Second stage



Introduction & System Model Guaranteed QoS SNR Margin Maximization

Instantaneous SNR

• The received signal at the *n*th destination is:

$$y_n = \mathbf{g}_n^{\mathcal{T}} \mathbf{t}_n + z_{d_n} = \mathbf{h}_n^{\mathcal{H}} \mathbf{w}_n s_n + z_n.$$

Instantaneous SNR at destination-n

$$\mathrm{SNR}_n = \frac{\sigma_{S_n}^2 |\mathbf{h}_n^{\mathcal{H}} \mathbf{w}_n|^2}{\sigma_R^2 \|\mathbf{G}_n^{1/2} \mathbf{w}_n\|^2 + \sigma_D^2}$$

• Let p_n be the total relay power allocated for user-n

$$p_n = \mathbb{E}\left[\|\mathbf{t}_n\|^2\right] = \mathbf{w}_n^{\mathcal{H}} \mathbf{D}_n \mathbf{w}_n.$$

Guaranteed QoS

Without Per-Relay Power Constraints

- Minimize the sum relay power with guaranteed QoS.
- Can be performed separately for each user



Introduction & System Model Guaranteed QoS SNR Margin Maximization

Without Per-Relay Power Constraints - Solutions

- Second-order cone programming (SOCP).
- Find *p_n* directly, determine the optimal beamformer **w**_{*n*} accordingly.

$$\mathcal{P}_n(\gamma_n) = \begin{cases} \underset{p_n}{\min \text{minimize}} & p_n \\ \text{subject to} & \sum_{i=1}^R \frac{a_{n,i}p_n}{b_{n,i}+p_n} \geq \gamma_n, \end{cases}$$

where $a_{n,i}$ and $b_{n,i}$ are parameters.

• Simple fixed point iteration

$$p_n^{(t+1)} = \frac{\gamma_n}{\sum_{i=1}^R \frac{a_{n,i}}{b_{n,i}+p_n^{(t)}}} \triangleq f_n(p_n^{(t)}).$$



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Introduction & System Model Guaranteed QoS SNR Margin Maximization

Convergence



• Fixed point iteration with different starting points.



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With Per-Relay Power Constraints

- Power constraint at each relay.
- Uniformly minimize the margin P_i/P_i^{max} , denoted as α .



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Optimization in Wireless Multi-relay Networks

Guaranteed QoS

Without Per-Relay Power Constraints - Solutions

- Second-order cone programming (SOCP).
- Study to dual problem:
 - The Lagrangian, the dual function, and the dual problem.
 - An equivalent virtual uplink channel to the dual problem



Introduction & System Model Guaranteed QoS SNR Margin Maximization

Convergence





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Power Consumption Comparison



 Red lines: without per-relay power constraints vs. Blue lines: with per-relay power constraints.



Introduction & System Model Guaranteed QoS SNR Margin Maximization

Sum Relay Power Constraint

• Jointly maximize the SNR-margin

 $\begin{array}{ll} \underset{\mathbf{w}_{1},\ldots,\mathbf{w}_{N}}{\text{maximize}} & \underset{n}{\min} \frac{\text{SNR}_{n}}{\gamma_{n}} \\ \text{subject to} & P_{\text{relay}} \leq P_{\text{relay}}^{\max}. \end{array}$

• Solutions: bisection method, modified fixed-point iteration to directly find the optimal solution:

$$\tilde{p}_n = \frac{\gamma_n}{\sum_{i=1}^R \frac{a_{n,i}}{b_{n,i} + p_n^{(t)}}}$$

then normalize the result

$$p_n^{(t+1)} = \frac{P_{\text{relay}}^{\max}}{\sum_{l=1}^N \tilde{p}_l}.$$

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Convergence



• Convergence of the modified fixed point iteration for each user and the corresponding SNR.



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Sum Relay Power Constraint

• Jointly maximize the SNR-margin

 $\begin{array}{ll} \underset{\mathbf{w}_{1},\ldots,\mathbf{w}_{N}}{\operatorname{maximize}} & \underset{n}{\min} \frac{\operatorname{SNR}_{n}}{\gamma_{n}} \\ \text{subject to} & P_{i} \leq P_{i}^{\max}, \ i = 1,\ldots,R. \end{array}$

• Solutions: bisection method, an iterative algorithm to directly find the optimal solution.



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Introduction & System Model Guaranteed QoS SNR Margin Maximization

Convergence





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Introduction & System Model Guaranteed QoS SNR Margin Maximization

SNR & Power Consumption Comparison



 Red lines: sum relay power constraint vs. Blue lines: per-relay power constraints.



Summary

- Distributed space-time coding: code design, power allocation, training and mismatched decoding.
- Distributed beamforming for multi-source multi-destination: power minimization with guaranteed QoS at the destinations, SNR margin maximization with power constraints at the relays.

QUESTIONS?

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Contributions

1	"Power Allocation and Error Performance of Distributed Unitary Space-Time Modulation in Wireless Relay Networks", to appear in IEEE Trans. on Veh. Tech.
2	"Channel Estimation and Performance of Mismatched Decoding in Wireless Relay Networks", <i>submitted to IEEE Trans. on Wireless Comm.</i>
3	"Resource Allocation in Wireless Multiuser Multi-relay Networks", in preparation.
4	"Distributed Beamforming in Relay-Assisted Multiuser Communications", <i>in Proc. IEEE ICC' 09</i> , Jun. 2009.
5	"A Novel Power Allocation Scheme for Distributed Space-Time Coding", in Proc. IEEE ICC' 09, Jun. 2009.
6	"Channel Estimation and Performance of Mismatched Decoding in Wireless Relay Networks", in Proc. IEEE ICC' 09, Jun. 2009.
7	"Power Allocation and Distributed Beamforming Optimization in Relay-Assisted Multiuser Communications", in Proc. IWCMC' 09, Jun. 2009.
8	"Distributed Unitary Space-Time Modulation in Partially Coherent and Noncoherent Relay Networks", in Proc. IPSPCS' 08, Dec. 2008.
9	"Distributed Beamforming in Multiuser Multi-relay Networks with Guaranteed QoS", <i>submitted to Globecom' 09.</i>
10	"SNR Maximization and Distributed Beamforming in Multiuser Multi-relay Networks", submmited to Globecom' 09.



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